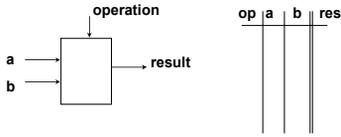
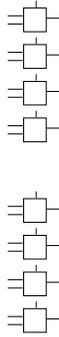


An ALU (arithmetic logic unit)

- Let's build an ALU to support the `andi` and `ori` instructions
 - we'll just build a 1 bit ALU, and use 32 of them



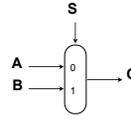
- Possible Implementation (sum-of-products):



1

Review: The Multiplexor

- Selects one of the inputs to be the output, based on a control input



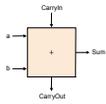
note: we call this a 2-input mux even though it has 3 inputs!

- Let's build our ALU using a MUX:

2

Different Implementations

- Not easy to decide the "best" way to build something
 - Don't want too many inputs to a single gate
 - Don't want to have to go through too many gates
 - for our purposes, ease of comprehension is important
- Let's look at a 1-bit ALU for addition:



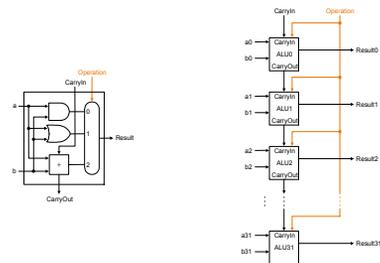
$$c_{out} = a b + a c_{in} + b c_{in}$$

$$sum = a \text{ xor } b \text{ xor } c_{in}$$

- How could we build a 1-bit ALU for add, and, and or?
- How could we build a 32-bit ALU?

3

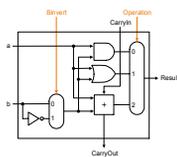
Building a 32 bit ALU



4

What about subtraction (a - b) ?

- Two's complement approach: just negate b and add.
- How do we negate?
- A very clever solution:



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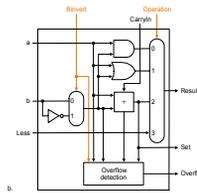
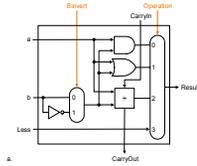
Tailoring the ALU to the MIPS

- Need to support the set-on-less-than instruction (`slt`)
 - remember: `slt` is an arithmetic instruction
 - produces a 1 if $rs < rt$ and 0 otherwise
 - use subtraction: $(a-b) < 0$ implies $a < b$
- Need to support test for equality (`beq $t5, $t6, $t7`)
 - use subtraction: $(a-b) = 0$ implies $a = b$

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Supporting slt

- Can we figure out the idea?

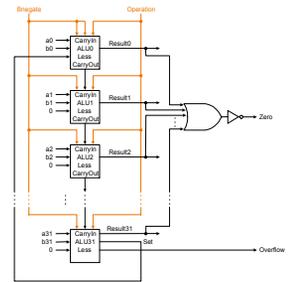


Test for equality

- Notice control lines:

000 = and
 001 = or
 010 = add
 110 = subtract
 111 = slt

**Note: zero is a 1 when the result is zero!*



8

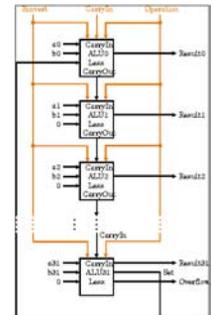
Conclusion

- We can build an ALU to support the MIPS instruction set
 - key idea: use multiplexor to select the output we want
 - we can efficiently perform subtraction using two's complement
 - we can replicate a 1-bit ALU to produce a 32-bit ALU
- Important points about hardware
 - all of the gates are always working
 - the speed of a gate is affected by the number of inputs to the gate
 - the speed of a circuit is affected by the number of gates in series (on the "critical path" or the "deepest level of logic")
- Our primary focus: comprehension, however,
 - Clever changes to organization can improve performance (similar to using better algorithms in software)
 - we'll look at two examples for addition and multiplication

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A 32-bit ALU

- A Ripple carry ALU
- Two bits decide operation
 - Add/Sub
 - AND
 - OR
 - LESS
- 1 bit decide add/sub operation
- A carry in bit
- Bit 31 generates overflow and set bit



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Problem: ripple carry adder is slow

- Is a 32-bit ALU as fast as a 1-bit ALU?
- Is there more than one way to do addition?
 - two extremes: ripple carry and sum-of-products

Can you see the ripple? How could you get rid of it?

$$\begin{aligned}
 c_1 &= b_0c_0 + a_0c_0 + a_0b_0 \\
 c_2 &= b_1c_1 + a_1c_1 + a_1b_1 \\
 c_3 &= b_2c_2 + a_2c_2 + a_2b_2 \\
 c_4 &= b_3c_3 + a_3c_3 + a_3b_3
 \end{aligned}$$

Not feasible! Why?

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Carry-look-ahead adder

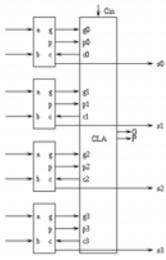
- An approach in-between our two extremes
- Motivation:
 - If we didn't know the value of carry-in, what could we do?
 - When would we always generate a carry? $g_i = a_i b_i$
 - When would we propagate the carry? $p_i = a_i + b_i$
- Did we get rid of the ripple?

$$\begin{aligned}
 c_1 &= g_0 + p_0c_0 \\
 c_2 &= g_1 + p_1c_1 \\
 c_3 &= g_2 + p_2c_2 \\
 c_4 &= g_3 + p_3c_3
 \end{aligned}$$

Feasible! Why?

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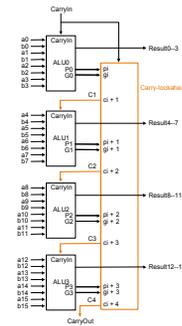
A 4-bit carry look-ahead adder



- Generate g and p term for each bit
- Use g's, p's and carry in to generate all C's
- Also use them to generate block G and P
- CLA principle can be used recursively

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Use principle to build bigger adders



- A 16 bit adder uses four 4-bit adders
- It takes block g and p terms and cin to generate block carry bits out
- Block carries are used to generate bit carries
 - could use ripple carry of 4-bit CLA adders
 - Better: use the CLA principle again!

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Delays in carry look-ahead adders

- 4-Bit case
 - Generation of g and p: 1 gate delay
 - Generation of carries (and G and P): 2 more gate delay
 - Generation of sum: 1 more gate delay
- 16-Bit case
 - Generation of g and p: 1 gate delay
 - Generation of block G and P: 2 more gate delay
 - Generation of block carries: 2 more gate delay
 - Generation of bit carries: 2 more gate delay
 - Generation of sum: 1 more gate delay
- 64-Bit case
 - 12 gate delays

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Multiplication

- More complicated than addition
 - accomplished via shifting and addition
- More time and more area
- Let's look at 3 versions based on grade school algorithm

$$\begin{array}{r} 01010010 \text{ (multiplicand)} \\ \times 01101101 \text{ (multiplier)} \\ \hline \end{array}$$

- Negative numbers: convert and multiply
- Use other better techniques like Booth's encoding

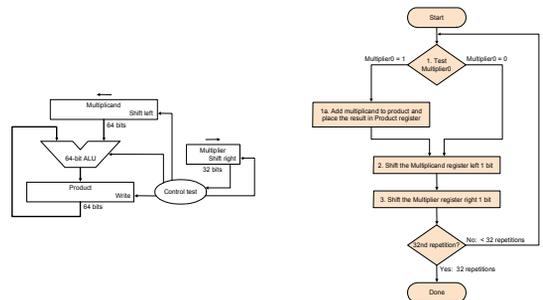
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Multiplication

$$\begin{array}{r} 01010010 \text{ (multiplicand)} \\ \times 01101101 \text{ (multiplier)} \\ \hline 00000000 \\ 01010010 \times 1 \\ \hline 01010010 \\ 00000000 \times 0 \\ \hline 001010010 \\ 010100100 \times 1 \\ \hline 0110011010 \\ 0101001000 \times 1 \\ \hline 10000101010 \\ 00000000000 \times 0 \\ \hline 010000101010 \\ 010100100000 \times 1 \\ \hline 0111001101010 \\ 010100100000 \times 1 \\ \hline 10001011101010 \\ 0000000000000 \times 0 \\ \hline 0010001011101010 \end{array}$$

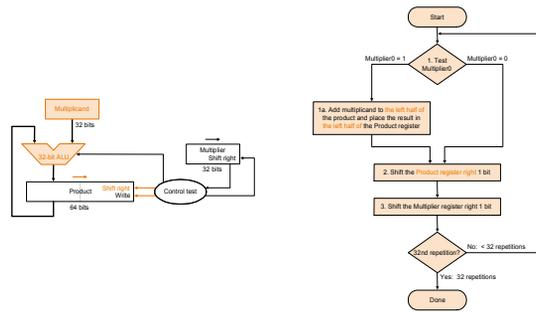
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Multiplication: Implementation



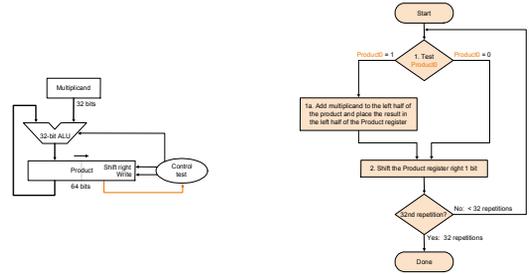
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Second Version



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Final Version



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Multiplication Example

Iteration	multi- plicand	Original algorithm	
		Step	Product
0	0010	Initial values	0000 0110
1	0010	1:0 ⇒ no operation	0000 0110
	0010	2: Shift right Product	0000 0011
2	0010	1a:1 ⇒ prod = Prod + Mcand	0010 0011
	0010	2: Shift right Product	0001 0001
3	0010	1a:1 ⇒ prod = Prod + Mcand	0011 0001
	0010	2: Shift right Product	0001 1000
4	0010	1:0 ⇒ no operation	0001 1000
	0010	2: Shift right Product	0000 1100

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Signed Multiplication

- Let Multiplier be $Q[n-1:0]$, multiplicand be $M[n-1:0]$
- Let $F = 0$ (shift flag)
- Let result $A[n-1:0] = 0 \dots 00$
- For $n-1$ steps do
 - $A[n-1:0] = A[n-1:0] + M[n-1:0] \times Q[0] / ^*$ add partial product $^*/$
 - $F <= F$.or. $(M[n-1] \text{ .and. } Q[0]) / ^*$ determine shift bit $^*/$
 - Shift A and Q with F, i.e.,
 - $A[n-2:0] = A[n-1:1]$; $A[n-1] = F$; $Q[n-1] = A[0]$; $Q[n-2:0] = Q[n-1:1]$
- Do the correction step
 - $A[n-1:0] = A[n-1:0] - M[n-1:0] \times Q[0] / ^*$ subtract partial product $^*/$
 - Shift A and Q while retaining $A[n-1]$
 - This works in all cases excepts when both operands are 10..00

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Booth's Encoding

- Numbers can be represented using three symbols, 1, 0, and -1
- Let us consider -1 in 8 bits
 - One representation is 1 1 1 1 1 1 1 1
 - Another possible one 0 0 0 0 0 0 -1
- Another example +14
 - One representation is 0 0 0 1 1 1 0
 - Another possible one 0 0 0 1 0 0 -1 0
- We do not explicitly store the sequence
- Look for transition from previous bit to next bit
 - 0 to 0 is 0; 0 to 1 is -1; 1 to 1 is 0; and 1 to 0 is 1
- Multiplication by 1, 0, and -1 can be easily done
- Add all partial results to get the final answer

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Using Booth's Encoding for Multiplication

- Convert a binary string in Booth's encoded string
 - Multiply by two bits at a time
 - For n bit by n -bit multiplication, $n/2$ partial product
 - Partial products are signed and obtained by multiplying the multiplicand by 0, +1, -1, +2, and -2 (all achieved by shift)
 - Add partial products to obtain the final result
 - Example, multiply 0111 (+7) by 1010 (-6)
 - Booths encoding of 1010 is -1 +1 -1 0
 - With 2-bit groupings, multiplication needs to be carried by -1 and -2
- $$\begin{array}{r}
 1\ 1\ 1\ 1\ 0\ 0\ 1\ 0 \\
 1\ 1\ 1\ 0\ 0\ 1\ 0\ 0
 \end{array}$$
- (multiplication by -2)
(multiplication by -1 and shift by 2 positions)
- Add the two partial products to get 11010110 (-42) as result

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