Linearization of Generator Current-State Space Model

These notes parallel sections 8.3-8.4 of VMAF.

We developed a state-space current model for the synchronous machine with the G-circuit represented (see notes on perunitization), and it was found to be:

$$\begin{bmatrix} v_{d} \\ -v_{F} \\ 0 \\ 0 \\ 0 \end{bmatrix} = -\begin{bmatrix} r & 0 & 0 & \omega L_{q} & \omega k M_{G} & \omega k M_{Q} \\ 0 & r_{F} & 0 & 0 & 0 & 0 \\ 0 & 0 & r_{D} & 0 & 0 & 0 \\ -\omega L_{d} & -\omega k M_{F} & -\omega k M_{D} & r & 0 & 0 \\ 0 & 0 & 0 & 0 & r_{G} & 0 \\ 0 & 0 & 0 & 0 & 0 & r_{Q} \end{bmatrix} \begin{bmatrix} i_{d} \\ i_{F} \\ i_{D} \\ i_{q} \\ i_{G} \\ i_{Q} \end{bmatrix}$$

$$- \begin{bmatrix} L_{d} & k M_{F} & k M_{D} & 0 & 0 & 0 \\ k M_{F} & L_{F} & M_{R} & 0 & 0 & 0 \\ k M_{D} & M_{R} & L_{D} & 0 & 0 & 0 \\ 0 & 0 & 0 & L_{q} & k M_{G} & k M_{Q} \\ 0 & 0 & 0 & k M_{G} & L_{G} & M_{Y} \\ 0 & 0 & 0 & k M_{Q} & M_{Y} & L_{Q} \end{bmatrix} \begin{bmatrix} i_{d} \\ i_{f} \\ i_{D} \\ i_{g} \\ i_{g} \end{bmatrix}$$

$$(L-1)$$

This was equation 4.74 in VMAF (in case the "dots" don't show up, note the right-hand-side vector of (L-1) is a vector of derivatives).

 $\lceil i \rceil$

The torque equation was also (see notes "TorqueEquation"):

$$\dot{\omega} = \frac{T_m}{\tau_j} + \left[\frac{-L_d}{3\tau_j}i_q \quad \frac{-kM_F}{3\tau_j}i_q \quad \frac{-kM_D}{3\tau_j}i_q \quad \frac{L_q}{3\tau_j}i_d \quad \frac{kM_G}{3\tau_j}i_d \quad \frac{kM_Q}{3\tau_j}i_d \quad \frac{-D}{\tau_j}\right] \begin{bmatrix} i_d \\ i_F \\ i_D \\ i_q \\ i_G \\ \frac{i_Q}{\omega} \end{bmatrix}$$
(L-2)

This is equation (4.101) in VMAF.

And finally, we had (equation (4.102) in VMAF):

$$\dot{\delta} = \omega - 1$$
 (L-3)

Note that the state variables are i_d , i_F , i_D , i_q , i_G , i_Q , ω , and δ , i.e.,

$$\underline{x} = \begin{vmatrix} i_d \\ i_F \\ i_D \\ i_Q \\ i_G \\ i_Q \\ \boldsymbol{\omega} \\ \boldsymbol{\delta} \end{vmatrix}$$
(L-4)

So we have 8 state variables and 8 equations.

We want to develop a model that will allow us to investigate the stability properties of the system represented by these equations using linear system theory, because the theory of linear systems is so powerfully well-developed, as we have seen in our previous work on eigenvalues, eigenvectors, mode shape, and participation factors.

However, to employ linear system theory, the state variables must only appear in our equations in the form of linear terms; that is, state variables may only appear as a term like a_ix_i where a_i is a constant.

Our equations have some state variables appearing as product terms of the form $x_i x_j$. For example, note in eq. 4.74 (L-1) the appearance of ω in the first matrix, which means that it will get multiplied by a current state variable to result in a term like $\omega L_q i_q$. Also, note that almost every term in the torque equation is the product of two current state variables.

So our current-state-space model is non-linear. The process of *linearization* is where we convert the equations so that they contain only linear terms. To do this, we will focus on each equation and convert each nonlinear term to a linear term.

Nothing is free. In order to linearize, we will need to select a "point" in the state-space. Then the price we will pay for the simplification is that the resulting model will only be valid for a region "close to" that selected point.

Our general strategy is based on the recognition that each of our equations in 4.74 look like

$$v = f(\underline{x}, \underline{\dot{x}}) \tag{L-5}$$

We will select our point of linearization as \underline{x}_0 and then we will develop a model based on the idea that

- Only changes in the state variables may occur (e.g., no changes in parameters);
- The changes in the state variables must be small.

We will denote the changes as:

$$\underline{x} = \underline{x}_0 + \underline{\Delta x} \tag{L-6}$$

Differentiating, we also have that:

$$\underline{\dot{x}} = \underline{\dot{x}}_0 + \underline{\Delta}\underline{\dot{x}} \tag{L-7a}$$

The term $\underline{\dot{x}}_0$ in eq. (L-7a) should not be understood as an attempt to take a set of derivatives on a set of constants (which would be zero), but rather a set of derivatives on \underline{x} that are subsequently evaluated at \underline{x}_0 . This understanding is emphasized by rewriting (L-7a) as:

$$\underline{\dot{x}} = \underline{\dot{x}}\Big|_{\underline{x}=\underline{x}_0} + \underline{\Delta}\underline{\dot{x}}$$
(L-7b)

The effect of changes in the state variables given by eqs. (L-6) and (L-7a) on eq. (L-5) can be expressed by substituting eqs. (L-6) and (L-7a) into eq. (L-5) and adjusting the right-hand-side of eq. (L-5) accordingly. That is:

$$v_0 + \Delta v = f(\underline{x}_0 + \Delta \underline{x}, \underline{\dot{x}}_0 + \Delta \dot{x})$$
 (L-8)

Therefore, every time we see a state variable, we will replace it with $x_0 + \Delta x$, and every time we see a state-variable derivative, we will replace it with $\dot{x}_0 + \Delta \dot{x}$.

Once this is done, however, the terms corresponding to products of state variables, i.e., x_1x_2 , will appear like

$$(x_{10} + \Delta x_1)(x_{20} + \Delta x_2)$$
 (L-9)

(Fortunately, we have no products of terms containing derivatives of state variables).

Eq. (L-9) is evaluated as:

 $(x_{10} + \Delta x_1)(x_{20} + \Delta x_2) = x_{10}x_{20} + x_{10}\Delta x_2 + x_{20}\Delta x_2 + \Delta x_1\Delta x_2$ (L-10) Here, for small deviations, $\Delta x_1\Delta x_2$ is negligible and therefore

$$(x_{10} + \Delta x_1)(x_{20} + \Delta x_2) \approx x_{10}x_{20} + x_{10}\Delta x_2 + x_{20}\Delta x_2$$
 (L-11)

Noting that

- the left-hand-side of eq. (L-11) gives the product term under the changed condition and
- the first term on the right-hand side gives the product term for the unchanged condition,

we see that the change in the product term is given by:

 $(x_{10} + \Delta x_1)(x_{20} + \Delta x_2) - x_{10}x_{20} = x_{10}\Delta x_2 + x_{20}\Delta x_1 \text{ (L-12)}$

We will repeatedly employ this strategy of substituting and then recognizing on the right-hand-side the "unchanged" expression.

Before we go through this procedure, it is worthwhile to note that the process of linearization comes about more formally through the Taylor series expansion (TSE) about a selected point.

Performing a TSE about the point \underline{x}_0 , we have:

$$f(\underline{x}_0 + \Delta \underline{x}) = f(\underline{x}_0) + f'(\underline{x}_0)\Delta \underline{x} + \frac{1}{2!}f''(\underline{x}_0)\Delta \underline{x}^2 + \dots (L-13)$$

Neglecting the higher-order terms with more than one derivative (justified by the idea that the higher order powers of $\Delta \underline{x}$ are very small), we have that

$$f(\underline{x}_0 + \Delta \underline{x}) \approx f(\underline{x}_0) + f'(\underline{x}_0) \Delta \underline{x}.$$
 (L-14)

For example, consider $f(x_1,x_2)=x_1x_2$, which is just a function with a single product term (which was what (L-12) was derived to address). The TSE of $f(x_1,x_2)=x_1x_2$ about the point $\underline{x}_0=[x_{10}, x_{20}]^T$ is:

$$f(\underline{x}_{0} + \Delta \underline{x}) \approx x_{10} x_{20} + \begin{bmatrix} x_{20} & x_{10} \end{bmatrix} \begin{bmatrix} \Delta x_{1} \\ \Delta x_{2} \end{bmatrix} = x_{10} x_{20} + x_{20} \Delta x_{1} + x_{10} \Delta x_{2}$$

where we see that

$$f(\underline{x}_{0} + \Delta \underline{x}) - f(\underline{x}_{0})$$

= $(x_{10}x_{20} + x_{20}\Delta x_{1} + x_{10}\Delta x_{2}) - (x_{10}x_{20})$
= $x_{20}\Delta x_{1} + x_{10}\Delta x_{2}$
which is in agreement with (L-12).

linearize one of them, for purposes of illustration. We take the first voltage equation, the one for v_d, lifted out of eq. 4.74 (or L-1 above), repeated below for convenience (done in 3 lines via 8.26 of VMAF): $v_d = -ri_d - \omega L_q i_q - \omega k M_G i_G - \omega k M_Q i_Q - L_d \dot{i}_d - k M_F \dot{i}_F - k M_D \dot{i}_D$ (L-15)

Substituting (note VMAF nomenclature is $v_{d\Delta}$; we use Δv_d here).

The eight state equations must be linearized one at a time. We

 $\omega \leftarrow \omega_0 + \Delta \omega$ we have:

$$\begin{aligned} v_{d0} + \Delta v_{d} &= \\ -r(\dot{i}_{d0} + \Delta \dot{i}_{d}) \\ -(\omega_{0} + \Delta \omega) L_{q} (\dot{i}_{q0} + \Delta \dot{i}_{q}) \leftarrow \\ -(\omega_{0} + \Delta \omega) k M_{G} (\dot{i}_{G0} + \Delta \dot{i}_{G}) \leftarrow \\ -(\omega_{0} + \Delta \omega) k M_{Q} (\dot{i}_{Q0} + \Delta \dot{i}_{Q}) \leftarrow \\ -L_{d} (\dot{i}_{d0} + \Delta \dot{i}_{d}) - k M_{F} (\dot{i}_{F0} + \Delta \dot{i}_{F}) - k M_{D} (\dot{i}_{D0} + \Delta \dot{i}_{D}) \end{aligned}$$
(L-16)

where the "product terms" are indicated by the arrow " \leftarrow ".

Recall eq. (L-11) which is the linearized expression for products, repeated below for convenience:

 $(x_{10} + \Delta x_{1})(x_{20} + \Delta x_{2}) = x_{10}x_{20} + x_{10}\Delta x_{2} + x_{20}\Delta x_{2}$ We apply this equation to each of the product terms in eq. (L-16). $v_{d0} + \Delta v_{d} =$ $-r(i_{d0} + \Delta i_{d})$ $-L_{q}(\omega_{0}i_{q0} + \omega_{0}\Delta i_{q} + \Delta\omega i_{q0}) \leftarrow$ $-kM_{G}(\omega_{0}i_{G0} + \omega_{0}\Delta i_{G} + i_{G0}\Delta\omega) \leftarrow$ $-L_{M}(\omega_{0}i_{Q0} + \omega_{0}\Delta i_{Q} + i_{Q0}\Delta\omega) \leftarrow$ $-L_{d}(i_{d0} + \Delta i_{d}) - kM_{F}(i_{F0} + \Delta i_{F}) - kM_{D}(i_{D0} + \Delta i_{D})$ Now separate, on the right-hand-side, initial condition terms (with

zero subscript) from "deviation" terms (with
$$\Delta$$
 in front):
 $v_{d0} + \Delta v_d = \left[-ri_{d0} - L_q \omega_0 i_{q0} - kM_Q \omega_0 i_{Q0} - kM_G \omega_0 i_{G0} - L_d \dot{i}_{d0} - kM_F \dot{i}_{F0} - kM_D \dot{i}_{D0} \right]$
 $+ \left\{ -r\Delta i_d - L_q \left(\omega_0 \Delta i_q + \Delta \omega i_{q0} \right) - kM_Q \left(\omega_0 \Delta i_Q + i_{Q0} \Delta \omega \right) - kM_G \left(\omega_0 \Delta i_G + i_{G0} \Delta \omega \right) - L_d \Delta \dot{i}_d - kM_F \Delta \dot{i}_F - kM_D \Delta \dot{i}_D \right\}$
(L-18)

Recognizing that the term in square brackets is just v_{d0} , we subtract v_{d0} from both sides, leaving only what is inside the curly brackets:

$$\Delta v_{d} = -r\Delta i_{d} - L_{q} \left(\omega_{0}\Delta i_{q} + \Delta \omega i_{q0} \right) - kM_{Q} \left(\omega_{0}\Delta i_{Q} + i_{Q0}\Delta \omega \right) - kM_{G} \left(\omega_{0}\Delta i_{G} + i_{G0}\Delta \omega \right) - L_{d}\Delta \dot{i}_{d} - kM_{F}\Delta \dot{i}_{F} - kM_{D}\Delta \dot{i}_{D}$$
(L-19)

Now combining terms in $\Delta \omega$, we obtain:

$$\Delta v_d = -r\Delta i_d - L_q \omega_0 \Delta i_q - kM_Q \omega_0 \Delta i_Q - kM_G \omega_0 \Delta i_G - L_d \Delta i_d - kM_F \Delta i_F - kM_D \Delta i_D$$
 (L-20)
$$- (L_q i_{q0} + kM_Q i_{Q0} + kM_G i_{G0}) \Delta \omega$$

The last term in (L-20) is interesting. Let me show you why.

We have previously derived that

$$\begin{bmatrix} \lambda_{0} \\ \lambda_{d} \\ \lambda_{q} \\ \lambda_{q} \\ \lambda_{F} \\ \lambda_{G} \\ \lambda_{Q} \end{bmatrix} = \begin{bmatrix} L_{0} & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & L_{d} & 0 & kM_{F} & 0 & kM_{D} & 0 \\ 0 & 0 & L_{q} & 0 & kM_{G} & 0 & kM_{Q} \\ 0 & kM_{F} & 0 & L_{F} & 0 & M_{R} & 0 \\ 0 & 0 & kM_{G} & 0 & L_{G} & 0 & M_{Y} \\ 0 & kM_{D} & 0 & M_{R} & 0 & L_{D} & 0 \\ 0 & 0 & kM_{Q} & 0 & M_{Y} & 0 & L_{Q} \end{bmatrix} \begin{bmatrix} i_{0} \\ i_{d} \\ i_{q} \\ i_{F} \\ i_{G} \\ i_{D} \\ i_{Q} \end{bmatrix}$$
(4.20)

This is eq. 4.20 in VMAF. (We developed this equation in the notes on "mach_eqts." It was also used in the notes on per-unitization and elsewhere.)

From the above, we see that

$$\begin{aligned} \lambda_d &= L_d i_d + k M_F i_F + k M_D i_D \\ \lambda_q &= L_q i_q + k M_Q i_Q + k M_G i_G \end{aligned}$$

Noting the equation for λ_q , we see that the coefficient on $\Delta \omega$ in eq. (L-20) is λ_{q0} . Therefore, eq. (L-20) becomes: Δv_d

$$= -r\Delta i_d - L_q \omega_0 \Delta i_q - kM_Q \omega_0 \Delta i_Q$$

$$- kM_G \omega_0 \Delta i_G - L_d \Delta \dot{i}_d - kM_F \Delta \dot{i}_F - kM_D \Delta \dot{i}_D - \lambda_{q0} \Delta \omega$$
(L-21)

And this is our linearized equation for the v_d voltage equation.

Similar analysis applies for the other 7 state equations. Once this is done, we may form a matrix relation like that of eq. 8.36, pg. 286 in the text, and the matrices will be 8×8 , as shown in (L-22).

The resulting matrix relation (which includes the 6 voltage equations and the 2 inertial equations) can be written more compactly as:

 $\mathbf{v} = -\mathbf{K}\mathbf{x} - \mathbf{M}\dot{\mathbf{x}} \qquad (\text{eq. 8.37})$

where the " Δ " on the variables **v**, **x**, and **xdot** is implied, and nomenclature **v**, **K**, **M**, **x**, and **xdot** are defined in (L-22).

Then we may solve for $\underline{\dot{x}}$ according to:

$$\dot{\mathbf{x}} = -\mathbf{M}^{-1}\mathbf{K}\mathbf{x} - \mathbf{M}^{-1}\mathbf{v} \text{ pu} \qquad (\text{eq. 8.38})$$

which is in the form of

$$\dot{\mathbf{x}} = \mathbf{A}\mathbf{x} + \mathbf{B}\mathbf{u} \qquad (\text{eq. 8.39})$$

where

$$\mathbf{A} = -\mathbf{M}^{-1}\mathbf{K}$$
$$\mathbf{B} = -\mathbf{M}^{-1}$$

The VMAF text, p. 287, also shows that the matrix **M** is related to the matrix we called **L** in chapter 4, according to

$$\mathbf{M} = \begin{bmatrix} \mathbf{L} & 0 & 0 \\ 0 & -\tau_j & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

On p. 4 of our notes called "LoadEquations," and in VMAF, Section 4.13 (p. 124), we developed the matrix **L** as

$$\mathbf{L} = \begin{bmatrix} L_d & kM_F & kM_D & 0 & 0 & 0 \\ kM_F & L_F & M_R & 0 & 0 & 0 \\ kM_D & M_R & L_D & 0 & 0 & 0 \\ 0 & 0 & 0 & L_q & kM_G & kM_Q \\ 0 & 0 & 0 & kM_G & L_G & M_Y \\ 0 & 0 & 0 & kM_Q & M_Y & L_Q \end{bmatrix}$$

We easily see in (L-22) that **L** is the upper left 6×6 submatrix of **M**.

Final Exam Question #4: Derive the 7th equation in (L-22), i.e., linearize equation (L-2) (equation (4.101) in VMAF). Although this is done in a very cursory way at the top of p. 286 of VMAF, I expect you to show work which follows a procedure similar to what is done above for the v_d equation.

Note: Final Exam questions #1-#3 are found on p. 23 of the notes called "pssdamp." Again, all final exam questions posed are to be worked untimed, with open book/notes, but without interaction with anyone else including classmates.

Eigenanalysis of linearized one-machine to infinite bus system:

Section 8.4 provides a linearization for a one-machine infinite bus problem, similar to what we did in the notes called "LoadEquations" corresponding to Section 4.13 of VMAF. Here, the generator is connected to an infinite bus having voltage of V_{∞} through an impedance of R_e +jL_e (ω =1). The changes to (4.103) of VMAF necessary to model the load are again what are necessary to (L-22):

- Where you see r, replace it with $r + R_e$ and denote it with \hat{R}
- Where you see L_d , replace it with $L_d + L_e$ and denote it with \hat{L}_d
- Where you see L_q , replace it with $L_q + L_e$ and denote it with \hat{L}_q
- Where you see λ_{d0} , replace it with $\hat{\lambda}_{d0} = \lambda_{d0} + L_e i_{d0}$
- Where you see λ_q , replace it with $\hat{\lambda}_{q0} = \lambda_{q0} + L_e i_{q0}$
- Insert $-K\cos(\delta_0 \alpha)$ as element (1,8) (the v_d equation) of matrix **K**
- Insert $-Ksin(\delta_0 \alpha)$ as element (3,8) (the v_q equation) of matrix **K**

Note that:

- \circ speed deviation equation contains un-hatted values for L_d, L_q, λ_{d0} , and λ_{q0} ;
- the last two bullets represent the Park's transformation of a set of balanced voltages at the infinite bus, where K=√3 V_∞ (and is <u>not</u> the same K as used in the saturation notes).
- $\circ \alpha$ is the angle of the infinite bus phasor voltage V_{∞} .

These changes to (L-22) result in

Γο]		\hat{R}		0		0	$\omega_0 \hat{L}$	'q	$\omega_0 k M_G$	$\omega_0 \mathbf{k} M_Q$	$\hat{\lambda}_{q0}$	$-K\cos(\delta_0-\alpha)$] [;]
		0		r_{F}		0	0		0	0	0	0	
$\left \begin{array}{c} V_F\\ 0\end{array}\right $		0		0	i	r _D	0		0	0	0	0	$\left \begin{array}{c} \iota_F \\ i \end{array} \right $
		$-\omega_0$	\hat{L}_{d}	$-\omega_0 \mathbf{k} M$	$-\omega_0$	kM _D	Ŕ		0	0	$-\hat{\lambda}_{d0}$	$-K\sin(\delta_0-\alpha)$	$\left\ \frac{\frac{v_D}{i}}{i}\right\ $
0	= -	0		0		0	0		r_G	0	0	0	$\begin{vmatrix} i_q \\ i_c \end{vmatrix}$
0		0		0		0	0		0	r_Q	0	0	$\begin{vmatrix} i_G \\ i_C \end{vmatrix}$
$\overline{T_m}$		$\left \frac{\lambda_{q0} - 1}{2} \right $	$L_d i_{q0}$	$\frac{-L_{AD}i_{q0}}{2}$		$\frac{i_{AD}i_{q0}}{2}$	$\frac{-\lambda_{d0}+1}{2}$	$L_q i_{d0}$	$\frac{L_{AQ}i_{d0}}{2}$	$\frac{L_{AQ}i_{d0}}{2}$	-D	0	$\frac{\underline{v}}{\omega}$
				5 0		5 0	5 0		0	3 0	-1	0	$\lfloor \delta \rfloor$
								ĸ					2
		\hat{L}_d	kM_F	kM_D	0	0	0	0	0 $[i_d]$				
		kM_F	L_{F}	M_{X}	0	0	0	0	$0 \mid \dot{i}_d \mid$				
		kM_D	M_{X}	L_D	0	0	0	0	$0 \mid \dot{i}_d$				
	_	0	0	0	\hat{L}_q	kM_G	kM_Q	0	$0 \ \overline{\dot{i}_d} \ $				
		0	0	0	kM_G	L_{G}	M_{Y}	0	$0 \mid \dot{i}_d \mid$				
		0	0	0	kM_Q	M_{Y}	L_Q	0	$0 \left\ \underline{i}_d \right\ $				
		0	0	0	0	0	0	$- au_{j}$	$0 \dot{\omega}$				
		0	0	0	0	0	0	0	$1 \left[\dot{\delta} \right]$				
					M								

Observe that in the above per-unit formulation, we have replaced the respective mutual inductances in the torque equation with L_{AD} and L_{AQ} .

Example 8.2 then computes the **M** and **K** matrices using the loading conditions of Example 5.3, with D=0. The M matrix (which is independent of loading) is shown to be

	2.100	1.550	1.550					
	1.550	1.651	1.550		0		0	
	1.550	1.550	1.605					
				2.040	1.490	1.490		
$\mathbf{M} =$		0		1.490	1.76	1.490	0	
				1.490	1.490	1.526		
		•••••		 	Δ		-1786.9	0
		U			U		0	1
				•				_

Necessary computations to obtain the **K** matrix are shown on p. 291 of VMAF, from which the **K** matrix is expressed as

					1			
	0.0211	0	0	2.040	1.490	1.490	1.4304	-1.025
	0	0.0007	0	0	0	0	0	0
	0	0	0.0131	0	0	0	0	0
K _	-2.100	-1.550	-1.550	0.0211	0	0	-1.0396	-1.397
N =	0	0	0	0	0.00584	0	0	0
	0	0	0	0	0	0.0540	0	0
	-0.014	-0.362	-0.362	-1.428	-0.790	-0.790	-D	0
	0	0	0	0	0	0	-1	0

The A matrix, given by $A = -M^{-1}K$, is then expressed as

		10	2		/	1			
	-36.080	0.412	14.177	-3488.32	-2547.84	-2547.84	-2445.93	1752.71]
	12.435	-4.684	77.202	1202.24	878.103	878.103	842.98	-604.06	
	22.835	4.125	-96.410	2207.74	1612.517	1612.517	1548.02	-1109.28	
A _	3613.82	2667.34	2667.34	-36.310	1.158	80.282	1789.02	2404.05	10-3
$\mathbf{A} =$	-416.28	-307.26	-307.26	4.183	-19.27	163.54	-206.08	-276.93	10
	-3122.11	-2304.41	-2304.41	31.369	17.68	-273.46	-1545.59	-2076.95	
	-0.0078	-0.2026	-0.2026	-0.7991	-0.4421	-0.4421	0.0	0.0	
	0.0	0.0	0.0	0.0	0	0.0	1000	0.0	

Example 8.3 then computes the eigenvalues from the A matrix (this can easily be done using Matlab) to obtain

$\lambda_{_{1}}$	=	-0.03594 + j0.9986	λ_5	=	-0.00132 + j0.03135
λ_{2}	=	-0.03594 - j0.9986	λ_6	=	-0.00132 - j0.03135
λ_3	=	-0.09928	λ_7	=	-0.000643
$\lambda_{_{4}}$	=	-0.28464	λ_8	=	-0.007113

A look at the imaginary parts indicates that corresponding modes have very low frequencies, e.g., eigenvalues 1 and 2 correspond to a 0.16Hz mode, and eigenvalues 5 and 6 correspond to a 0.005Hz mode. However, we are forgetting that time has been normalized, and so these eigenvalues are expressed in rad/(sec/t_{base})=(rad/sec)*t_{base}. To express the eigenvalues in units of rad/sec, we need to divide by t_{base}, which is equivalent to multiplying by 377. Multiplying these frequencies by 377, we obtain 60 Hz and 1.88Hz. Some comments:

- VMAF on p. 292 explain the 60Hz mode as being "the 60-Hz component injected into the rotor circuits to **balance** the MMF caused by the stator DC currents." I add that we observe this mode as a result of the fact that we used the full IEEE 2.2 model where we still represent stator transients, i.e., we have NOT made the assumption in this model that stator transients are negligible.
- The 1.88 Hz mode is a little higher than the modes we were studying in the eigenanalysis of our project system, which were 0.28-0.30 Hz and 0.41-0.42 Hz. However, we looked only at these modes because they were the ones giving us trouble (they are *interarea* modes). If we would have inspected the entire list of modes that SSAT computed for us, we would have found many modes in the 1-2.5Hz range. For example, here are a few I picked out:

161	-2.3566	7.5618	1.2035	29.75	Base.SSA	Base Scenari	io	No fault		
					Dominant State :	13 : NAVAJO 226.0 :	0 : : 1	: GENROU :	:	Angle
119	-2.4101	13.2805	2.1137	17.86	Base.SSA	Base Scenari	io	No fault		
					Dominant State :	77 : JOHN DAY13.8 :	0 : : 1	: ESDC2A : TE	:	1
144	-2.4529	10.3648	1.6496	23.03	Base.SSA	Base Scenari	io	No fault		
					Dominant State :	47 : OWENS G 11.5 :	0 : : 1	: GENROU :	:	Speed
206	-2.4778	2.9667	0.4722	64.10	Base.SSA	Base Scenari	io	No fault		
					Dominant State :	140 : LITEHIPE20.0 :	0 : : 1	: ESDC2A : TE	:	1

The vertical box identifies the frequency, in Hz, of four modes, were we observe frequencies at 1.2 Hz (dominated by Navajo angle state), 2.1 Hz (dominated by an exciter state), 1.65 Hz (dominated by Owens speed state). The 2.1 and 1.65 Hz modes are referred to as "local modes." If we looked at their participation factors, we would find significant participation from states associated with only those units.

<u>Point</u>: Interarea modes typically have frequencies 0.1-1.0Hz. Local modes typically have frequencies above 1.0Hz.

Last comments on linearized analysis:

• Section 8.5 of VMAF shows how to linearize the flux-linkage state-space model; it is satisfying that on p. 298 it indicates that eigenanalysis of the resulting **A**-matrix provides the same eigenvalues as we obtained using the current state-space model.

Section 8.6, called "State Matrix for Multimachine System," provides a very good articulation of how to build the A-matrix for a multimachine system. This section is analogous to the treatment for time-domain simulation given in Chapter 7 where we saw we must express all dynamic models on a common reference frame. Section 8.6 also shows how to account for nonlinear loads such as a ZIP or exponential models (again, analogous to Chapter 7), an issue of particular interest since the nonlinear equations must be linearized. It is interesting to note at the beginning of this section that it says,

"The basis for this procedure is the work done in reference [4] in the development of small-signal stability analysis software, which is now available as part of Powertech Labs' DSATools package."

Reference [4] is

[4] Kundur, P., Rogers G.J., Wong D.Y., Arabi, S., and Wang, L. Small Signal Stability Analysis Program: Version 3.1, Vol. 1 Final Report, EPRI TR-101850-V1R1, May 1994.

which is an EPRI report that came out a few years after the 1990 paper on the website.

Kundur, Rogers, Wong, Wang, and Lauby, "A comprehensive computer program package for small signal stability analysis of power systems," IEEE Transactions on Power Systems, Vol. 5, No. 4, Nov., 1990.

Although there are several major commercial-grade time-domain simulation packages used in the industry (PSLF, PSS\E, DSAtools, and PowerWorld), there is only one commercial-grade eigenanalysis package, which is SSAT.